

# RADIO COMMUNICATION APPARATUS

## *CROSS REFERENCE TO RELATED APPLICATION*

This application is based upon and claims the benefit of priority from the prior  
5 Japanese Patent Application No. 2002-253798, filed on August 30, 2002, the entire  
contents of which are incorporated herein by reference.

## *BACKGROUND OF THE INVENTION*

### *1. Field of the Invention*

10 The present invention relates to a radio communication apparatus and, in particular,  
to a transmitter-receiver of a base station or a mobile station in a mobile communication  
system.

### *2. Description of the Related Art*

15 In recent years, highly advanced digital transmission technologies and digital signal  
processing technologies have been applied to many radio transmission systems and radio  
applied equipment. In particular, there are strong demands for mobile communication  
system terminals and other equipment using these technologies to be adaptable to a variety  
of multimedia and to improve their added values.

20 It is, accordingly, highly required that antenna systems of such radio transmission  
systems and radio applied equipment not only have a desired radio transmission path or  
wireless zone but also be adaptable to a variation in the transmission characteristics of a radio  
transmission path and secure high channel quality with a low cost. Moreover, weight  
reduction, miniaturization, power saving, and higher reliability are also required.

25 The following three techniques are conventional techniques for abating channel

quality deterioration by coping with fading and a multipath phenomenon that may occur in such a radio transmission path:

First technique: Maximum SN ratio combining is performed so as to attain a maximum effective SN ratio by generating, as shown in Fig. 5, an arriving wave as a subject of demodulation and a signal judgment as the sum of products of arriving waves R1 and R2 arriving at two elements 91-1 and 91-2. The elements 91-1, 91-2 constitute an array antenna 90 and are arranged at such a short distance  $d$  as to produce sufficiently strong spatial correlation and weights W1 and W2, which are generated according to a prescribed algorithm (e.g., an algorithm that attains prescribed digital beam forming).

Second technique: Diversity reception is performed by generating, as shown in Fig. 6, an arriving wave as a subject of demodulation and a signal judgment as the sum of products of arriving waves  $r1$  and  $r2$  that are received in parallel via two antennas 101-1 and 101-2. The antennas 101-1, 101-2 are arranged at such a long distance  $d$  as to produce sufficiently weak spatial correlation and prescribed weights  $w1$  and  $w2$ .

Third technique: An arriving wave as a subject of demodulation and a signal judgment are generated as shown in Fig. 7 as the sum of products of arriving waves that are received via elements of each of a plurality of  $n$  array antennas 90-1 to 90- $n$  that are arranged at such a long distance  $d$  as to produce sufficiently weak spatial correlation and weights that are generated for the respective elements according to the above-mentioned algorithm. The SN ratio is increased in accordance with the above-mentioned digital beam forming and diversity gain.

Note that in the first technique, even in the case where the arrival direction of a desired arriving wave is unknown or may vary, a null point can be formed in the arrival direction of a disturbing (undesired) wave according to the above-mentioned algorithm.

Therefore, the channel quality can be prevented from being degraded greatly even when the

DU ratio is very low or is varying largely. However, a variation in the level of an arriving wave is not always sufficiently suppressed because of strong spatial correlation between the elements 91-1 and 91-2.

Using the second technique, it is possible to suppress a variation in the level of an arriving wave with reliability because spatial correlation between the antennas 101-1 and 101-2 is sufficiently weak. However, a main lobe cannot be formed in the arrival direction of a desired arriving wave, and neither can a null point in the arrival direction of a disturbing wave.

Therefore, in a radio transmission system in which complex multipaths are formed, in particular, good channel quality cannot be obtained with stability. In using this technique for transmission, since unintended radiation of a transmission wave occurs via a grating lobe that is formed in a direction in which a receiving end is not located, the technique is not adaptable to a desired zone configuration, channel allocation, and frequency allocation, which makes difficult to use this technique in practice.

In the third technique, a disturbing wave or an interference wave cannot always sufficiently be suppressed when the DU ratio of an arriving wave at one of the array antennas 90-1 to 90-n is very low. Bit errors may occur in the process of judging signals that are obtained by performing diversity combining of such arriving waves.

#### *SUMMARY OF THE INVENTION*

An object of the present invention is to provide a radio communication apparatus capable of suppressing an interference wave with reliability, adapting to the configuration and the characteristics of a radio transmission path, and capable of compensating for a variation in the transmission characteristics of the radio transmission path with stability and accuracy.

Another object of the invention is to attain in parallel suppression of a disturbing

(undesired) wave by beam and null forming and diversity reception by combining processing even in the case where the level of an arriving wave or a disturbing wave may vary largely.

Another object of the invention is to reduce processing amount necessary for beam forming.

5 Another object of the invention is to enable the radio communication apparatus to be adapted to different distances and various arrangements of antennas.

Still another object of the invention is to simplify the configuration of the radio communication apparatus and enhance the responsiveness and the reliability thereof.

Yet another object of the invention is to form a full-duplex radio transmission path  
10 of a frequency division scheme with reliability without impairing suppression of a disturbing wave by beam and null forming and diversity reception by combining processing.

A further object of the invention is to enhance the performance, added values, and total reliability of a radio transmission system or a radio applied system to which the invention is applied and to keep a high and stable total reliability.

15 One or some of the above objects are attained by a radio communication apparatus which applies a particular set of weights in common to a plurality of array antennas, the particular set of weights being to be applied to an array antenna that has received an arriving wave with maximum channel quality, and which combines arriving waves received with the array antennas.

20 One or some of the above objects are attained by a radio communication apparatus characterized in that the particular set of weights are such values as to allow each of the array antennas to function as an adaptive null-forming array antenna.

One or some of the above objects are attained by a radio communication apparatus which applies to the plurality of array antennas a particular set of weights which allow one of  
25 the array antennas to have a main lobe and a null point in the arrival directions of a desired

wave and of a disturbing wave, respectively, the one of the array antennas having received the desired wave and the disturbing wave as arriving waves with good channel quality.

One or some of the above objects are attained by a radio communication apparatus which corrects weights to values such that the one of the array antennas has a main lobe and  
5 a null point in the directions of a desired wave and a disturbing wave, respectively, and applies the corrected weights in common to the plurality of array antennas.

One or some of the above objects are attained by a radio communication apparatus characterized in that each of the array antennas is composed of elements which are arranged on a same virtual line or plane parallel to each position of the array antennas.

10 One or some of the above objects are attained by a radio communication apparatus which further comprises section provided on a feed line of each of all or part of the array antennas, for using a set of weights for transmission of a transmission wave via the feed line. The set of weights are obtained by correcting the particular set of weights in accordance with a frequency difference between the transmission wave and an arriving wave.

15 One or some of the above objects are attained by a radio communication apparatus in which each of all or part of the array antennas is paired with a transmission array antenna, and which additionally has a section for applying the particular set of weights to a feed line of each transmission array antenna.

One or some of the above objects are attained by a radio communication apparatus  
20 which adapts to aerial beam forming antennas instead of the array antennas and applies a set of reactances to be loaded on elements of the aerial beam forming antennas instead of the particular set of weights.

One or some of the above objects are attained by a radio communication apparatus which adapts to a plurality of adaptive beam forming array antennas independently  
25 performing beam forming instead of the array antennas, and corrects a set of weights to the

above-mentioned particular set of weights. The set of weights are updated for each of adaptive beam forming array antennas, realizing the independently performed beam forming.

One or some of the above objects are attained by a radio communication apparatus which adapts to a plurality of adaptive null-forming array antennas independently performing  
5 beam forming instead of the array antennas, and corrects a set of weights to the above-mentioned particular set of weights. The set of weights are updated for each of adaptive null-forming array antennas, realizing the independently performed beam forming.

The above radio communication apparatus will be described below more specifically.

In a first radio communication apparatus according to the invention, a channel  
10 quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of array antennas. A computing section calculates, for elements of each of the array antennas, a set of weights which are such values as to allow each of the array antennas to function as an adaptive beam forming array antenna. A weight setting section selects a particular set of weights from the calculated sets of weights and applies the particular set of  
15 weights in common to the plurality of array antennas, the particular set of weights being to be applied to an array antenna that has received an arriving wave with maximum channel quality as monitored by the channel quality monitoring section. A combining section combines the arriving waves that have been received with the array antennas to which the particular set of weights are applied.

20 In this radio communication apparatus, the array antennas can continuously perform desired accurate beam forming, using a particular set of weights which are to be applied to an array antenna that has received an arriving wave with highest channel quality among the array antennas.

This realizes suppression of a disturbing wave arriving together with a desired wave,  
25 stable diversity reception through the beam forming, and combining processing performed by

the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each array antenna may vary largely.

In a second radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of array antennas. A computing section calculates arrival angles of a desired wave and a disturbing wave as the arrival waves for each of the array antennas. A weight setting section selects, from the calculated arrival angles, arrival angles of a desired wave and a disturbing wave as arrival waves with good channel quality as monitored by the channel quality monitoring section, and it applies a particular set of weights in common to the array antennas, the particular set of weights being such values as to allow each of the plurality of array antennas to have a main lobe and a null point in directions of the selected arrival angles of the desired wave and the disturbing wave, respectively. A combining section combines the arriving waves that have been received with the array antennas to which the particular set of weights are applied.

This radio communication apparatus can obtain a particular set of weights to be applied to the array antennas without calculating unnecessary sets of weights that are not applied to any array antennas, as long as the arrival angles of a desired wave and a disturbing wave are calculated with desired accuracy.

This realizes reduction in processing amounts required for beam forming of the array antennas as well as suppression of a disturbing wave arriving together with a desired wave, stable diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each array antenna may vary largely.

In a third radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a

plurality of array antennas. A computing section calculates arrival angles of a desired wave and a disturbing wave as the arriving waves and a set of weights for each of the array antennas, the set of weights being such values as to allow each of the array antennas to function as an adaptive null-forming array antenna. A weight setting section selects, from the calculated arrival angles, arrival angles of a desired wave and a disturbing wave as arrival waves with good channel quality as monitored by the channel quality monitoring section, corrects one of the calculated sets of weights to such values as to allow an array antenna to form a main lobe and a null point in directions of the selected arrival angles of the desired wave and the disturbing wave, respectively, the array antenna having received an arriving wave with maximum channel quality as measured by the channel quality monitoring section. The weight setting section then applies the corrected set of weights in common to all the array antennas. A combining section combines the arriving waves that have been received with the array antennas to which the corrected set of weights are applied.

This radio communication apparatus applies, to the array antennas, particular sets of weights which are obtained by correcting sets of weights calculated for the respective array antennas by the computing section in the above-described manner.

This makes the radio communication apparatus adaptable to different distances and various arrangements of the array antennas, and this also achieves suppression of a disturbing wave, stable diversity reception through predetermined beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each array antenna may vary largely.

In a fourth radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of aerial beam forming antennas. A computing section calculates a set of reactances for elements of each of the aerial beam forming antennas, the set of reactances being loaded



on elements of each of the aerial beam forming antennas. A reactance setting section selects from the calculated sets of reactances a particular set of reactances which are to be loaded on an aerial beam forming antenna having received an arriving wave with maximum channel quality as monitored by the channel quality monitoring section, and applies the particular set of reactances in common to all the aerial beam forming antennas. A combining section combines the arriving waves that have been received with the aerial beam forming antennas on which the particular set of reactances are loaded.

In this radio communication apparatus, the aerial beam forming antennas can continuously perform desired accurate beam forming using a particular set of reactances which are to be loaded on an array antenna that has received an arriving wave with highest channel quality among the aerial beam forming antennas.

This realizes suppression of a disturbing wave arriving with a desired wave, stable diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each array antenna may vary largely.

In a fifth radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of adaptive beam forming array antennas. A weight setting section selects, from sets of weights being loaded on the respective adaptive beam forming array antennas, a particular set of weights which are to be applied to an adaptive beam forming array antenna that has received an arriving wave with maximum channel quality as monitored by the channel quality monitoring section. It then applies the particular set of weights in common to the adaptive beam forming array antennas as corrected values. A combining section combines the arriving waves that have been received with the adaptive beam forming array antennas.

In this radio communication apparatus, the adaptive beam forming array antennas can continuously perform desired accurate beam forming simultaneously, using a particular set of weights, without impairing updates of weights to be applied to the adaptive beam forming array antennas to preferable values. The particular set of weights are to be applied to an adaptive beam forming array antenna that has received an arriving wave with highest channel quality.

This realizes suppression of a disturbing wave arriving with a desired wave, stable diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each adaptive beam forming array antenna may vary largely.

In a sixth radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of array antennas. A computing section calculates, for elements of each of the array antennas, a set of weights that are such values as to allow each of the array antennas to function as an adaptive null-forming array antenna. A weight setting section selects, from the calculated sets of weights, a particular set of weights that are to be applied to an array antenna that has received an arriving wave with maximum channel quality as monitored by the channel quality monitoring section. It then applies the particular set of weights in common to the array antennas. A combining section combines the arriving waves that have been received with the array antennas to which the particular set of weights are applied.

In this radio communication apparatus, the array antennas can continuously form null points in the arrival directions of disturbing waves using a particular set of weights which are to be applied to an array antenna that has received an arriving wave with highest channel quality among the array antennas.

This realizes suppression of a disturbing wave arriving with a desired wave, stable

diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each array antenna may vary largely.

In a seventh radio communication apparatus according to the invention, each of the  
5 array antennas is composed of elements which are arranged on a same virtual line or plane parallel to each position of the plurality of array antennas.

This radio communication apparatus is capable of applying a set of weights in common to the array antennas as long as an arriving wave arrives at each array antenna in the form of a plane wave.

10 Therefore, it is possible to simplify the configuration of the radio communication apparatus and enhance the responsiveness and the reliability thereof, compared with the case where it is necessary to calculate sets of weights for each of the array antennas.

In an eighth radio communication apparatus according to the invention, provided is/are feeding section(s) on feed line(s) of all or part of the array antennas, for employing a set  
15 of weights for transmission of a transmission wave via the feed line. The set of weights are obtained by correcting the particular set of weights in accordance with a frequency difference between the transmission wave and an arriving wave.

In this radio communication apparatus, all or part of the array antennas can be used for both reception of an arriving wave and transmission of a transmission wave even in the  
20 case where the arriving wave and the transmission wave have different frequencies from each other.

This makes it possible to form a full-duplex radio transmission path of a frequency division scheme with high reliability without impairing suppression of a disturbing wave by the above-described beam forming and diversity reception.

25 In a ninth radio communication apparatus according to the invention, each of all or

part of the array antennas is paired with a transmission array antenna which is used for transmitting a transmission wave having a different frequency than that of an arriving wave. Also provided is/ are feeding section(s) for applying the particular set of weights to a feed line of the transmission array antenna.

5           In this radio communication apparatus, the computing section and the weight setting section can be used for both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

          This makes it possible to form a full-duplex radio transmission path of a frequency  
10   division scheme with high reliability without impairing suppression of a disturbing wave by the above-described beam forming and diversity reception.

          In a 10th radio communication apparatus according to the invention, a channel  
quality monitoring section monitors channel quality of each of arriving waves that arrive at a  
plurality of aerial beam forming antennas. A computing section calculates, for elements of  
15   each of the aerial beam forming antennas, a set of reactances which are to be loaded on each  
of the aerial beam forming antennas and are such values as to allow each aerial beam forming  
antenna to function as an adaptive null-forming array antenna. A reactance setting section  
selects, from the calculated sets of reactances, a particular set of reactances which are to be  
loaded on an aerial beam forming antenna that has received an arriving wave with maximum  
20   channel quality as monitored by the channel quality monitoring section. It applies the  
particular set of reactances in common to all the aerial beam forming antennas. A combining  
section combines the arriving waves that have been received with the aerial beam forming  
antennas on which the particular set of reactances are loaded.

          In this radio communication apparatus, the aerial beam forming antennas can  
25   continuously form null points with accuracy in the arrival directions of disturbing waves using

a particular set of reactances which are to be loaded on an array antenna that has received an arriving wave with highest channel quality among the aerial beam forming antennas.

This realizes suppression of a disturbing wave arriving with a desired wave, stable diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each aerial beam forming antenna may vary largely.

In an 11th radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of aerial beam forming antennas. A computing section calculates arrival angles of a desired wave and a disturbing wave as the arrival waves for each of the aerial beam forming antennas. A reactance setting section selects, from the calculated arrival angles, arrival angles of a desired wave and a disturbing wave which are arrival waves with good channel quality as monitored by the channel quality monitoring section. It applies a particular set of reactances in common to all the aerial beam forming antennas. The particular set of reactances are such values to allow each of the plurality of aerial beam forming antennas to have a main lobe and a null point in directions of the selected arrival angles of the desired wave and the disturbing wave, respectively. A combining section combines the arriving waves that have been received with the aerial beam forming antennas on which the particular set of reactances are loaded.

This radio communication apparatus can obtain a particular set of reactances to be loaded on the aerial beam forming antennas without calculating unnecessary sets of reactances that are not applied to any aerial beam forming antennas, as long as the arrival angles of a desired wave and a disturbing wave are calculated with desired accuracy

This realizes reduction in processing amounts required for beam forming of the aerial beam forming antennas as well as suppression of a disturbing wave arriving with a

desired wave, stable diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each aerial beam forming antenna may vary largely.

In a 12th radio communication apparatus according to the invention, a channel  
5 quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of aerial beam forming antennas. A computing section calculates a set of reactances and arrival angles of a desired wave and a disturbing wave as the arriving waves for each of the aerial beam forming antennas. The set of reactances are such values as to allow each of the aerial beam forming antennas to function as an adaptive null-forming array antenna. A  
10 reactance setting section selects, from the calculated arrival angles, arrival angles of a desired wave and a disturbing wave that are arrival waves with good channel quality as monitored by the channel quality monitoring section. And it corrects one of the calculated sets of reactances to such values as to allow an aerial beam forming antenna to have a main lobe and a null point in directions of the selected arrival angles of the desired wave and the disturbing  
15 wave, respectively. The aerial beam forming antenna has received an arriving wave with maximum channel quality as measured by the channel quality monitoring section. The reactance setting section applies the corrected set of reactances in common to all the aerial beam forming antennas. A combining section combines the arriving waves that have been received with the aerial beam forming antennas on which the corrected set of reactances are  
20 loaded.

This radio communication apparatus applies, to each of the aerial beam forming antennas, a particular set of reactances which are obtained by correcting sets of reactances calculated for a selected aerial beam forming antenna by the computing section.

This makes the radio communication apparatus adaptable to different distances and  
25 various arrangements of the aerial beam forming antennas, and this also achieves

suppression of a disturbing wave, stable diversity reception through predetermined beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each aerial beam forming antenna may vary largely.

5           In a 13th radio communication apparatus according to the invention, each of the aerial beam forming antennas is composed of elements which are arranged on a same virtual line or plane parallel to each position of the plurality of aerial beam forming antennas.

          This radio communication apparatus is capable of applying a set of reactances in common to the aerial beam forming antennas as long as an arriving wave arrives at each  
10   aerial beam forming antenna in the form of a plane wave.

          Therefore, it is possible to simplify the configuration of the radio communication apparatus and enhance the responsiveness and the reliability thereof, compared with the case where it is necessary to calculate sets of reactances for each of the aerial beam forming antennas.

15           In a 14th radio communication apparatus according to the invention, provided is/are feeding section(s) on feed line(s) of all or part of the aerial beam forming antennas. The feeding section(s) uses(s) a set of reactances for transmission of a transmission wave via the feed line. The set of reactances are obtained by correcting the particular set of reactances in accordance with a frequency difference between the transmission wave and an arriving wave.

20           In this radio communication apparatus, all or part of the aerial beam forming antennas can be used for both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

          This makes it possible to form a full-duplex radio transmission path of a frequency  
25   division scheme with high reliability without impairing suppression of a disturbing wave by

the above-described beam forming and diversity reception.

In a 15th radio communication apparatus according to the invention, each of all or part of the aerial beam forming antennas is paired with a transmission aerial beam forming antenna which is to be used for transmitting a transmission wave with a different frequency from that of an arriving wave. Further provided is/are feeding section(s) for applying the particular set of reactances to feeding lines of the aerial beam forming antennas.

In this radio communication apparatus, the computing section and the reactance setting section can be used for both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

This makes it possible to form a full-duplex radio transmission path of a frequency division scheme with high reliability without impairing suppression of a disturbing wave by the above-described beam forming and diversity reception.

In a 16th radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of adaptive null-forming array antennas. A weight setting section selects a particular set of weights from sets of weights which are to be loaded on the respective adaptive null-forming array antennas. The particular set of weights are applied to an adaptive null-forming array antenna that has received an arriving wave with maximum channel quality as monitored by the channel quality monitoring section. The weight setting section applies the particular set of weights as corrected values in common to the adaptive null-forming array antennas. A combining section combines the arriving waves that have been received with the adaptive null-forming array antennas.

In this radio communication apparatus, the adaptive null-forming array antennas can simultaneously form null points in the arrival directions of disturbing waves with accuracy,



using a particular set of weights without impairing updates of sets of weights to be applied to the adaptive null-forming array antennas to preferable values. The particular set of weights are to be applied to an adaptive null-forming array antenna that has received an arriving wave with highest channel quality.

5 This realizes suppression of a disturbing wave arriving with a desired wave, stable diversity reception through the beam forming, and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each adaptive null-forming array antenna may vary largely.

In a 17th radio communication apparatus according to the invention, a channel  
10 quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of adaptive null-forming array antennas. A computing section calculates arrival angles of a desired wave and a disturbing wave as the arriving waves for each of the plurality of adaptive null-forming array antennas. A weight setting section selects, from the calculated arrival angles, arrival angles of a desired wave and a disturbing wave that are arrival  
15 waves having good channel quality as monitored by the channel quality monitoring section, and it applies a particular set of weights in common to all the adaptive null-forming array antennas as corrected values. The particular set of weights refer to such values as to allow each of the plurality of adaptive null-forming antennas to have a main lobe and a null point in directions of the selected arrival angles of the desired wave and the disturbing wave,  
20 respectively. A combining section combines the arriving waves that have been received with the adaptive null-forming array antennas.

This radio communication apparatus is able to obtain a particular set of weights to be applied to the adaptive null-forming antennas without calculating unnecessary sets of weights that are not applied to any adaptive null-forming array antennas and without  
25 impairing updates of sets of weights to be applied to the adaptive null-forming array

antennas to preferable values, as long as the arrival angles of a desired wave and a disturbing wave are calculated with desired accuracy.

Therefore, it is possible to reduce processing amounts necessary for beam forming of the adaptive null-forming array antennas as well as to effectively suppress a disturbing wave arriving with a desired wave, and perform stable diversity reception through the beam forming and combining processing performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each adaptive null-forming array antenna may vary largely.

In an 18th radio communication apparatus according to the invention, a channel quality monitoring section monitors channel quality of each of arriving waves that arrive at a plurality of adaptive null-forming array antennas. A computing section calculates arrival angles of a desired wave and a disturbing wave as the arriving waves for each of the adaptive null-forming array antennas. A weight setting section selects, from the calculated arrival angles, arrival angles of a desired wave and a disturbing wave that are arrival waves with good channel quality as monitored by the channel quality monitoring section, and it corrects a set of weights being to be applied to an adaptive null-forming antenna to such values as to allow the adaptive null-forming array antenna to have a main lobe and a null point in directions of the selected arrival angles of the desired wave and the disturbing wave, respectively. The adaptive null-forming antenna is the one that has received an arriving wave with maximum channel quality as measured by the channel quality monitoring section. The weight setting section applies the corrected set of weights in common to all the adaptive null-forming array antennas. A combining section combines the arriving waves that have been received with the adaptive null-forming array antennas.

This radio communication apparatus applies to the adaptive null-forming array antennas a particular set of weights which are obtained by correcting sets of weights

calculated for the adaptive null-forming array antennas by the computing section.

This makes the radio communication apparatus adaptable to different distances and various arrangements of the adaptive null-forming array antennas, and this also realizes suppression of a disturbing wave, stable diversity reception through predetermined beam forming, and combining processing that is performed by the combining section even in the case where the level of an arriving wave or a disturbing wave that arrives at each adaptive null-forming array antenna may largely vary.

In a 19th radio communication apparatus according to the invention, each of the adaptive null-forming array antennas is composed of elements that are arranged on a same virtual line or plane parallel to each position of the plurality of adaptive null-forming array antennas.

In this radio communication apparatus, it is possible to apply a set of weights in common to the adaptive null-forming array antennas as long as an arriving wave arrives at every adaptive null-forming array antenna in the form of a plane wave.

Therefore, it is possible to simplify the configuration of the radio communication apparatus and enhance the responsiveness and the reliability thereof, compared with the case where it is necessary to calculate sets of weights for each of the adaptive null-forming array antennas.

In a 20th radio communication apparatus according to the invention, further provided is/ are feeding section(s) on feed line(s) of all or part of the adaptive beam forming array antennas. The feeding section(s) employ(s) a set of weights for transmission of a transmission wave via the feed line(s). The set of weights are obtained by correcting the particular set of weights in accordance with a frequency difference between the transmission wave and an arriving wave.

In this radio communication apparatus, all or part of the adaptive beam forming array

antennas can be used for both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

5 This realizes secure formation of a full-duplex radio transmission path of a frequency division scheme without impairing suppression of a disturbing wave by the above-described beam forming and diversity reception.

10 In a 21st radio communication apparatus according to the invention, further provided is/ are feeding section(s) on feed line(s) of all or part of the adaptive null-forming array antennas. The feeding section(s) employ(s) a set of weights for transmission of a transmission wave via the feed line(s). The set of weights are obtained by correcting the particular set of weights in accordance with a frequency difference between the transmission wave and an arriving wave.

15 In this radio communication apparatus, all or part of the adaptive null-forming array antennas can be used for both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

This realizes secure formation of a full-duplex radio transmission path of a frequency division scheme without impairing suppression of a disturbing wave by the above-described beam forming and diversity reception.

20 In a 22nd radio communication apparatus according to the invention, all or part of the adaptive beam forming array antennas is/are paired with transmission array antenna(s) which is/are to be used for transmitting a transmission wave with a different frequency than that of an arriving wave. Further provided is/are feeding section(s) for applying the particular set of weights to feeding line(s) of the transmission array antenna(s).

25 In this radio communication apparatus, the weight setting section can be used for

both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

This achieves secure formation of a full-duplex radio transmission path of a frequency division scheme via the adaptive beam forming array antennas and transmission array antennas without impairing suppression of a disturbing wave by the above-described beam forming and diversity reception.

In a 23rd radio communication apparatus according to the invention, all or part of the adaptive null-forming array antennas is/are paired with transmission array antenna(s) which is/are to be used for transmitting a transmission wave with a different frequency than that of an arriving wave. Further provided is/are feeding section(s) for applying the particular set of weights to feeding line(s) of the transmission array antenna(s).

In this radio communication apparatus, the computing section and the weight setting section can be used for both reception of an arriving wave and transmission of a transmission wave even in the case where the arriving wave and the transmission wave have different frequencies from each other.

This realizes secure formation of a full-duplex radio transmission path of a frequency division scheme without impairing suppression of a disturbing wave by the above-described beam forming, and diversity reception.

#### *BRIEF DESCRIPTION OF THE DRAWINGS*

The nature, principle, and utility of the invention will become more apparent from the following detailed description when read in conjunction with the accompanying drawings in which like parts are designated by identical reference numbers, in which:

Fig. 1 shows first to third embodiments of the invention;

Fig. 2 shows an example of improved channel quality achieved by the first embodiment;

Fig. 3 shows the detailed configuration of an antenna duplexer;

Fig. 4 shows modifications of the first to third embodiments of the invention;

5 Fig. 5 shows the configuration of a first conventional antenna system;

Fig. 6 shows the configuration of a second conventional antenna system; and

Fig. 7 shows the configuration of a third conventional antenna system.

### *DESCRIPTION OF THE PREFERRED EMBODIMENTS*

10 Embodiments of the present invention will be hereinafter described with reference to the drawings.

Fig. 1 shows first to third embodiments of the invention.

As shown in Fig. 1, each of array antennas 50-1 to 50-n is composed of a plurality p of elements 51-11 to 51-1p, ..., 51-n1 to 51-np that have the same structure and  
15 arrangement in a direction in which an arriving wave (for the sake of simplicity, it is assumed here that an arriving wave is a plane wave) may arrive. And the array antennas 50-1 to 50-n are arranged at such a long distance as to produce sufficiently weak mutual spatial correlation. In the following description, common items in association with the array antennas 50-1 to 50-n will be given a character "C", signifying that it can be any of the  
20 reference numerals 50-1 to 50-n and 51-1 to 51-n.

The feeding points of the elements 51-C1 to 51-Cp are connected to the inputs of A/D converters (A/D) 52-C1 to 52-Cp. The outputs of the A/D converters 52-C1 to 52-Cp are connected to one inputs of multipliers 53-C1 to 53-Cp, respectively. The outputs of the multipliers 53-C1 to 53-Cp are connected to the respective inputs of each of a signal  
25 processing part 54-C and an adder ( $\Sigma$ ) 55-C. The output of the adder 55-C is connected to

one input of a multiplier 56-C and a corresponding input of an auxiliary signal processing part 57. The plurality n of outputs of the auxiliary signal processing part 57 are connected to the other inputs of the multipliers 56-1 to 56-n, respectively. The outputs of the multipliers 56-1 to 56-n are connected to the respective inputs of an adder 58. The output of the adder 58 is connected to a corresponding input of the auxiliary signal processing part 57. A baseband signal representing an arriving wave is obtained at the output of the adder 58.

One output of the signal processing part 54-C is connected to a corresponding input of a feed controlling part 59. The plurality p of outputs of the feed controlling part 59 are connected to the other inputs of the multipliers 53-C1 to 53-Cp, respectively. The other output of the signal processing part 54-C is connected to a corresponding input of a channel quality determining part 60. The output of the channel quality determining part 60 is connected to the control input of the feed controlling part 59.

The operation of the first embodiment of the invention will be described below with reference to Fig. 1.

The signal processing part 54-C performs the following processing:

(1) Calculates a correlation matrix  $RC_{yy}$  that is given by the following equation for a column vector  $YC(t)$  having, as elements, arriving waves  $yC1$  to  $yCn$  that arrive at the elements 51-C1 to 51-Cp in parallel and are supplied in parallel from the feeding points of the elements 51-C1 to 51-Cp via the A/D converters 52-C1 to 52-Cp and the multipliers 53-C1 to 53-Cp and a row vector  $Y(t)^H$  having, as elements, complex numbers  $yC1^*$  to  $yCp^*$  that are conjugate to the arriving waves  $yC1$  to  $yCn$ :

$$RC_{yy} = E[YC(t) \cdot YC(t)^H] \quad \dots (1)$$

(2) Calculates channel quality QC as evaluation values  $eC(t)$  that are given by the following equation for a known signal (e.g., a pilot signal)  $rC(t)$  that was received as arriving waves  $yC1$  to  $yCp$  and the above-described column vector  $YC(t)$ :

$$eC(t) = |YC(t) - rC(t)|^2 \quad \dots (2)$$

(3) Calculates a reception error  $rC_{er}$  ( $= E[eC(t) \cdot rC(t)^*]$ ), and collectively calculates, for the reception error  $rC_{er}$ ,  $n$  weights to be supplied in parallel to the inputs (referred to above as “the other inputs”) of the multipliers 53-C1 to 53-Cp to form a weight vector  $WC_{opt}$  that is  
5 given by the following equation and has the  $n$  weights arranged in time-series order,:

$$WC_{opt} = RC_{yy}^{-1} \cdot rC_{er} \quad \dots (3)$$

On the other hand, the channel quality determining part 60 determines maximum channel quality  $Q_{max}$  among the channel quality  $Q_1$  to  $Q_n$ , and determines a particular one of the array antennas 50-1 to 50-n that has received an arriving wave with the maximum  
10 channel quality  $Q_{max}$ .

The feed controlling part 59 supplies, in parallel, the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np with a particular combination of weights calculated for the particular array antenna among the combinations of weights calculated collectively as the elements of the weight vectors  $WC_{opt}$  for the array antennas 50-1 to 50-n.

15 The adder 55-C combines the arriving waves that arrive at the elements 51-C1 to 51-Cp and are supplied in parallel from their feeding points via the A/D converters 52-C1 to 52-Cp and the multipliers 53-C1 to 53-Cp.

Therefore, an arriving wave obtained by suppressing a disturbing (undesired) wave contained in the arriving wave arriving at the array antenna 50-C is obtained at the output of  
20 the adder 55-C.

The multipliers 56-1 to 56-n and the adder 58 produce a baseband signal (mentioned above) as the sum of products of weights that are supplied in parallel from the auxiliary signal processing part 57 and the arriving waves.

With the above operation, even if the DU ratio of an arriving wave or the level of a  
25 desired wave contained therein is very low, the array antennas 50-1 to 50-n can perform



beam forming stably by using the same weights as for a particular array antenna that provides maximum channel quality of a received arriving wave.

Under such beam forming disturbing waves are suppressed, and the multipliers 56-1 to 56-n and the adder 58 perform diversity reception in cooperation with each other under the initiative of the auxiliary signal processing part 57.

Therefore, according to this embodiment, a higher, more stable channel quality can be attained (for example, as indicated by a solid line in Fig. 2) than in a conventional example (indicated by a broken line in Fig. 2).

In this embodiment, the signal processing part 54-C generates weights so that the array antenna 50-C can serve as an adaptive beam forming array antenna that forms a main lobe in a direction in which a desired wave arrives and copes with direction changes.

However, the weights may be such values as to allow the array antenna 50-C to serve as an adaptive null-forming array antenna that also forms a null point in the arrival direction of a disturbing wave.

In this embodiment, each of the array antennas 50-1 to 50-n is a combination of the elements that have the same structure and arrangement in a direction in which an arriving wave may arrive.

However, the invention is not limited to such a configuration. For example, where the elements constituting each of the array antennas 50-1 to 50-n do not have a common arrangement in the above-mentioned direction or have different structures, the feed controlling part 59 or a component capable of replacing it may calculate, with prescribed accuracy at prescribed response speed, weights that absorb the differences between the directions or structures of the elements of the array antennas 50-1 to 50-n.

The second embodiment of the invention will be hereinafter described.

This embodiment is characterized in the following procedure of processing that is

performed by the signal processing part 54-C, the feed controlling part 59, and the channel quality determining part 60.

The operation of the second embodiment of the invention will be described below with reference to Fig. 1.

5       The feed controlling part 59 has a weight table in which all combinations of azimuth angles  $\theta_d$  indicating directions in which a desired wave may arrive and azimuth angles  $\theta_u$  indicating directions in which a disturbing wave may arrive and a combination of suitable weights to be applied, in common, to the array antennas 50-1 to 50-n for each of the above combinations.

10       The signal processing part 54-C not only calculates channel quality QC (described above) but also calculates, in the following manner, an arrival angle  $\theta_u$  indicating a direction in which a maximum-level disturbing wave arrives and an arrival angle  $\theta_d$  of a desired wave:

15       - Calculates an arrival angle  $\theta_d$  of a desired wave contained in arriving waves  $y_{C1}$  to  $y_{Cp}$  according to an arrival direction estimation algorithm such as a high-resolution algorithm.

20       - Determines, as a maximum eigenvalue of a correlation matrix  $RC_{yy}$  (described above), a maximum level among levels of disturbing wave contained in the arriving waves  $y_{C1}$  to  $y_{Cp}$ , and calculates an arrival angle  $\theta_u$  indicating the arrival direction of the disturbing wave having the maximum level as an eigenvector of the correlation matrix  $RC_{yy}$  corresponding to the maximum eigenvalue.

The channel quality determining part 60 determines maximum channel quality  $Q_{max}$  among the pieces of channel quality  $Q_1$  to  $Q_n$ , and determines a particular array antenna that has received an arriving wave with the channel quality  $Q_{max}$ .

25       The feed controlling part 59 selects particular arrival angles  $\theta_d$  and  $\theta_u$  that have been calculated for the particular array antenna among the arrival angles  $\theta_d$  and  $\theta_u$  calculated

by the signal processing parts 54-1 to 54-n.

Further, the feed controlling part 59 supplies the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np with particular weights that are registered in the weight table as corresponding to the combination of the particular arrival angles  $\Theta_d$  and  $\Theta_u$ .

5 With the above operation, even if the DU ratio of an arriving wave or the level of a desired wave contained therein is very low, the array antennas 50-1 to 50-n can perform beam forming stably by using such weights as to allow formation of both of a main lobe and a null point in directions in which a desired wave and a disturbing wave contained in an arriving wave having maximum channel quality arrive among arriving waves received in parallel by the  
10 array antennas 50-1 to 50-n.

Therefore, this embodiment achieves suppression of a disturbing wave which occurs due to all the array antennas 50-1 to 50-n's functioning as adaptive null-forming array antennas as well as predetermined diversity reception performed by the multipliers 56-1 to 56-n and the adder 58 under the initiative of the auxiliary signal processing part 57.

15 In this embodiment, supplied to the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np is a particular combination of weights that are such values as to allow the array antennas 50-1 to 50-n to function as adaptive null-forming array antennas.

However, the particular combination of weights may be such a combination of weights as to allow the array antennas 50-1 to 50-n to function as mere adaptive beam  
20 forming array antennas that do not form a null point in the arrival direction of a disturbing wave.

In this embodiment, a combination of weights, which allows formation of a null point in the arrival direction of only one with a maximum level among disturbing waves arriving at the array antennas 50-1 to 50-n in parallel, is supplied in common to the multipliers 53-11  
25 to 53-1p, ..., 53-n1 to 53-np.

However, the invention is not limited to such a configuration. For example, where a plurality of disturbing waves may arrive at the array antennas 50-1 to 50-n in parallel, a combination of weights that allows formation of a null point in an arrival angle  $\Theta_u$  that is calculated in such a manner that eigenvalues of correlation matrices  $RC_{yy}$  corresponding to the respective disturbing waves are used as weights and that is given as an weighted average of eigenvectors of the correlation matrices  $RC_{yy}$  corresponding to the respective eigenvalues may be used in a similar manner.

In this embodiment, all the combinations of weights to be supplied, in common, to the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np are stored in the weight table in advance.

However, the invention is not limited to such a configuration. For example, where a certain reduction in response speed is allowable, the hardware scale and the power consumption may be reduced by calculating a combination of weights to be supplied, in common, to the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np by arithmetic operations (suitable for the structures, arrangements, etc. of the array antennas 50-1 to 50-n) that are performed when necessary in accordance with particular arrival angles  $\Theta_d$  and  $\Theta_u$ .

In this embodiment, a particular combination of weights selected by the feed controlling part 59 from the combinations of weights that are registered in the weight table in advance is supplied, in common, to the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np.

However, the invention is not limited to such a configuration. For example, different combinations of weights that are obtained by performing, on combinations of weights calculated by the respective signal processing parts 54-1 to 54-n, corrections of compressing deviations of arrival angles of one or both of desired waves and disturbing waves contained in arriving waves arriving in parallel at the array antennas 50-1 to 50-n from arrival angles calculated for an arriving wave arriving with maximum channel quality among those arriving waves may be supplied to the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np. This is to

flexibly adapt to physical or geographical distances between the installation sites of the array antennas 50-1 to 50-n and differences between the structures and the element arrangements of the array antennas 50-1 to 50-n.

Such corrections may not necessarily be performed on all the combinations of weights calculated by the respective signal processing parts 54-1 to 54-n. One of simplification of the configuration, power saving, and increase in response speed may be attained by, for example, supplying a common combination of weights to multipliers downstream of array antennas that are regarded as having no physical or geographical distances between their installation sites or differences between their structures and element arrangements.

The third embodiment of the invention will be hereinafter described.

An antenna duplexer (DUP) 61-1 is disposed between the feeding points of the elements 51-11 to 51-1p of the array antenna 50-1 and the inputs of the A/D converters 52-11 to 52-1p. A transmission wave signal (for the sake of simplicity, it is assumed that the transmission wave signal whose occupied band is different from the band of an arriving wave) is supplied to the plurality n of transmission inputs of the antenna duplexer 61-1 via a branching filter 62. Corresponding outputs of the feed controlling part 59 are connected to the plurality p of weight inputs of the antenna duplexer 61-1.

As shown in Fig. 3, the antenna duplexer 61-1 is composed of the following components:

- Duplexing parts 61D-11 to 61D-1p that serve for impedance matching with the feeding points of the elements 51-11 to 51-1p, have a sharp attenuation pole in the occupied bandwidth of the transmission wave signal, and passes, to the A/D converters 52-11 to 52-1p, occupied-bandwidth components of arriving waves that are supplied from the feeding points in parallel.

- Transmission wave processing parts 61TP-11 to 61TP-1p that multiply the transmission wave signal supplied via the branching filter 62 by weights supplied from the feed controlling part 59, and supply resulting generated transmission waves to the feeding points of the elements 51-11 to 51-1p.

5           The operation of the third embodiment of the invention will be described below with reference to Fig. 1.

This embodiment is characterized by the following processing that is performed by the antenna duplexer 61-1 and the feed controlling part 59.

10           As in the case of the first or second embodiment, the feed controlling part 59 supplies the multipliers 53-C1 to 53-Cp with a combination of weights to be used for causing the array antennas 50-1 to 50-n to function as adaptive beam forming array antennas or adaptive null-forming array antennas.

In parallel with the processing of supplying such a combination of weights to the multipliers 53-C1 to 53-Cp, the feed controlling part 59 performs the following processing:

15           - Every time a combination of weights to be supplied to the multipliers 53-C1 to 53-Cp is determined, calculates such a combination of weights as to allow the array antenna 50-1 to function as an adaptive beam forming array antenna or an adaptive null-forming array antenna in the occupied band of a transmission wave signal on the basis of the thus-determined combination of weights and a frequency difference between an arriving  
20           wave and the transmission wave.

- Supplies the calculated combination of weights to the transmission wave processing parts 61TP-11 to 61TP-1p.

That is, a transmitting part that is composed of the branching filter 62 and the transmission wave processing parts 61TP-11 to 61TP-1p is roughly coupled with the A/D  
25           converters 52-11 to 52-1p and a receiving part downstream thereof on the frequency axis by

the above-mentioned attenuation pole.

Further, the array antenna 50-1 can serve for both transmission and reception that are suitable for a desired channel arrangement and frequency allocation because the transmission wave processing parts 61TP-11 to 61TP-1p are supplied with a combination of weights in accordance with a frequency difference between a transmission wave and an arriving wave.

Therefore, according to this embodiment, disturbing waves contained in arriving waves are suppressed and diversity reception is attained with the use of the adder 58 because the array antennas 50-1 to 50-n function as adaptive beam forming array antennas or adaptive null-forming array antennas. Further, this embodiment prevents unintended radiation of a transmission wave to regions other than a wireless zone that is formed according to the directivity of the array antenna 50-1.

In this embodiment, only the array antenna 50-1 serves for both transmission and reception. However, the invention is not limited to such a configuration. For example, diversity transmission may be attained by the following configuration:

- Antenna duplexers corresponding to the above-described antenna duplexer 61-1 are additionally provided for respective desired ones of the array antennas 50-1 to 50-n.
- The branching filter 62 distributes a transmission wave signal to those antenna duplexers in parallel.
- The feed controlling part 59 supplies a combination of weights that is suitable for a frequency difference between the transmission wave and an arriving wave to the transmission wave processing parts of those antenna duplexers in parallel.

In this embodiment, a combination of weights to be supplied to the transmission wave processing parts 61TP-11 to 61TP-1p is generated by correcting a combination of weights to be supplied to the multipliers 53-11 to 53-1p in accordance with a frequency

difference between a transmission wave and an arriving wave.

However, the invention is not limited to such a configuration. For example, the above-mentioned frequency difference may be absorbed by an array antenna that is provided separately from the array antenna 50-1 and serves for only transmission, and a common combination of weights may be supplied in parallel to the transmission wave processing parts 61TP-11 to 61TP-1p and the multipliers 53-11 to 53-1p.

In each of the above embodiments, the combination of weights to be supplied to the multipliers 53-11 to 53-1p, ..., 53-n1 to 53-np in parallel is updated by the feed controlling part 59.

However, the invention is not limited to such a configuration. For example, where the combination of weights is updated for each array antenna, it can be configured to immediately cope with a rapid variation in the transmission characteristic of a radio transmission path and to quickly stabilize channel quality of a newly formed radio channel by correcting an updated combination of weights at a prescribed frequency.

In each of the above embodiments, adaptive beam forming array antennas or adaptive null-forming array antennas are formed as a result of the operation that the characteristics of the feed lines of the array antennas 50-1 to 50-n are updated according to the adaptive algorithm.

However, the invention is not limited to radio transmission systems using such array antennas 50-1 to 50-n. For example, the invention is also applicable to equipment of an antenna system such as a radio terminal that accesses a mobile communication system or a wireless LAN that is required to satisfy not only a low cost and low power consumption but also a wide dynamic range.

As shown in Fig. 4, a low cost, low power consumption, and a wide dynamic range may be realized by using, instead of the array antennas 50-1 to 50-n, RESAA(Reactively



Steered Adaptive Array) antennas (aerial beam forming antennas) in which only a particular element is supplied with power directly and the reactance to be loaded on each of elements disposed in the vicinity of the particular element is varied according to a prescribed algorithm.

Where the equipment of an antenna system incorporates the aerial beam forming antennas therein for serving for both transmission and reception, for example, separate elements may be provided therein and directly used for transmission and reception, respectively. Alternatively, a certain element used for both transmission and reception may be provided separately (as a stack or the like) or it may be used in a suitable form for a desired multiple access scheme or frequency allocation, be invalidated electronically when necessary.

Where the above-mentioned RESAA antennas are used, each signal processing part cannot directly refer to arriving waves that arrive at the respective elements other than the above-mentioned particular element. When arriving waves that arrive at the respective elements of an array antenna are combined together in a radio frequency range or an intermediate frequency band, the signal processing part 54-C cannot directly refer to the arriving waves that arrive at the respective elements.

However, in either case, the signal processing part 54-C calculates, according to the following procedure, the value of reactance with which each element of the RESAA antenna is to be loaded or a set of weights to be applied to the respective elements of the array antenna:

- Temporarily changes reactance values or weights for the respective elements during a period when a known pilot signal or the like arrives as an arriving wave.

- Calculates a preferable combination of reactance values or weights according to an algorithm for obtaining such unknowns as to establish a known mathematical relationship between the changed reactance values or weights and arriving waves that are received in a state that the changed reactance values or weights are used, and that all the elements are in

cooperation with one another.

In each of the above embodiments, a same combination of weights is applied to all the array antennas 50-1 to 50-n.

However, the invention is not limited to such a configuration. For example,  
5 installation sites of the array antennas 50-1 to 50-n may be allocated in various ways by applying a same combination of weights only to part of the array antennas 50-1 to 50-n that are not very distant from each other geographically or physically.

As for array antennas installed at distant sites, a combination of weights that are corrected in accordance with all or part of the following items may be applied:

- 10       – Positions relative to the other array antennas.
- An area where a transmitting end of an arriving wave may exist.
- A desired zone configuration.

In each of the above embodiments, the invention is applied to the antenna system of a radio terminal that accesses a mobile communication system or a wireless LAN.

15       However, the invention is not limited to radio transmission systems and radio communication systems such as mobile communication system and wireless LANs. For example, the invention can also be applied to measuring instruments and a variety of radio applied systems such as navigation systems, positioning systems, and ranging systems.

In each of the above embodiments, the multipliers 56-1 to 56-n and the adder 58  
20 perform maximum SN ratio combining under the initiative of the auxiliary signal processing part 57.

However, the invention is not limited to the case of performing such maximum SN ratio combining. Any type of combining such as co-phase combining, minimum dispersion combining, or notch detection type MID combining may be employed as long as a desired  
25 diversity effect is attained that is suitable for all or part of a multiple access scheme, a

modulation scheme, a zone configuration, and channel allocation and the transmission characteristic of a radio transmission path.

In each of the above embodiments, the operations such as multiplications for attaining desired beam forming or null-forming are performed in the baseband.

5           Alternatively, equivalent operations may be performed in the radio frequency band or the intermediate frequency band as long as they adapt to the equipment configuration and a frequency band, channel allocation, a zone configuration, a multiple access scheme, a modulation scheme, etc.

10           The invention is not limited to the above embodiments and various modifications may be made without departing from the spirit and scope of the invention. Any improvement may be made in part or all of the components.